

### Weight Selection for Pattern Control of Paraboloidal Reflector Antennas with Reconfigurable Rim Scattering



Virginia Wireless (a)

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See <a href="https://www.faculty.ece.vt.edu/swe/raim/">https://www.faculty.ece.vt.edu/swe/raim/</a>



### Motivation

- Interference from satellites is a long-standing problem for radio astronomy
- Traditional methods of mitigation include avoidance (i.e., not observing) and post-observation editing (i.e., deleting afflicted portions of observations).
- Emerging and planned satellite systems will consist of thousands of satellites
- It is unclear how traditional methods of interference mitigation will fare once these systems are fully deployed.





### System Model

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- We are investigating the use of reconfigurable rim scattering where a large reflector is equipped with a passive reflectarray that creates a spatial null in the direction of the interferer
- Example We will assume D = 18 m paraboloidal reflector operating at 1.5 GHz
- The outer 0.5 m of the reflector surface consists of 2756 contiguous reconfigurable segments.
- Each segment is a square flat plate conformal to the paraboloidal surface having side length 0 . 5λ ( i.e. , an area  $\Delta s = 0 . 25\lambda^2$  ).



# Modifying the Received Electric Field

• Received electric field:

$$\mathbf{E}^{s}(\psi) = \underbrace{\mathbf{E}^{s}_{f}(\psi)}_{f} + \underbrace{\mathbf{E}^{s}_{r}(\psi)}_{r} \underbrace{\text{Due to the reconfigurable}}_{\text{portion of the dish}}$$

Due to the fixed portion of the dish

o To cancel sidelobe at angle  $\psi_o$  we desire:

$$E_f^{s,co}(\psi_o) + E_r^{s,co}(\psi_o) = 0$$

$$E_r^{s,co}(\psi_o) = \mathbf{e}_{\psi}^T \mathbf{w}$$

How do we determine the weights **w** ?





### Determining the Weights

• The optimal weights are

$$\mathbf{w}_{opt} = -E_f^{s,co}(\psi_o) \frac{\mathbf{e}_{\psi}^*}{||\mathbf{e}_{\psi}||_2^2}$$

• Disallowing amplification (i.e., phase change only):

$$\mathbf{w}_{gp} = \min_{\mathbf{w} \in \mathcal{C}^N} \quad \left\| E_f^{s,co}(\psi_0) + \mathbf{e}_{\psi}^T \mathbf{w} \right\|_2^2$$
  
s.t.  $|w_i| = 1$   $i = 1, 2, \dots N$ 





## Determining the weight vector **w**

#### Bad news

- The cost function is non-convex and extremely complex.
- The search space is large 2756 complex variables for 18m dish with 2m reconfigurable rim.

#### Good news

- There appear to be a large number of very well-performing local minima
- Finding a good solution is not particularly difficult require no more than 10,000 iterations





### Infinite Phase Quantization



Quantized Weights

Algorithm: Simulated
Annealing

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- Quantizing phase to *K* values requires a search over space of *K*<sup>2748</sup> vectors
- There appear to be a very large number (at least 1000's) of (good) local minima
- RIGHT: Gain achieved for  $1^o \le \psi \le 3^o$



### Main Lobe Variation – Infinite Quantization



### Main Lobe Variation – Quantized Weights

 Mainlobe constraints can also be applied in the quantized weight case

 Search is more complex, thus there is more variation







## Initial "Real-Time" Closed-loop Simulation



 Initial work assumed we know the pattern – open-loop search for weights

 Without knowledge of the pattern we must use a real-time (closed-loop) search

 LEFT: Gain of reconfigurable rim antenna as source moves from 0 to 3 degrees. Binary weights; INR = 0dB; integration over 1000 samples, angular velocity = 0.8deg/sec



### Conclusions

 In this work we have shown that with phase-only reconfigurable patches placed on the rim of a parabolic reflectarray antenna nulls can be placed in the pattern

- Despite the large search space and complex cost function, good weight vectors can be found relatively quickly (thousands of iterations)
- Infinite phase quantization yields arbitrarily low sidelobe gains.
- Binary and quarternary weights appear to be more limited, but also perform well
  - For our example antenna, only the first sidelobe is problematic.
- Initial results for "closed-loop" solution provided promising results.





# QUESTIONS?